

SC5306B 1 MHz to 3.9 GHz RF Downconverter Core Module

Datasheet

SC5306B SPECIFICATIONS

Definition of Terms

The following terms are used throughout this datasheet to define specific conditions:

Specification (spec)

Defines guaranteed performance of a calibrated instrument under the following conditions:

- 3 hours storage at room temperature (standardized to 25 °C) followed by 30 minutes minimum warm-up operation.
- Specified environmental conditions are met within the specified operating temperature range of 0 °C to 40 °C unless otherwise noted.
- Recommended calibration intervals are used.

Typical data (typ)

When used with <, > or in a range, defines performance met by approximately 80% of all instruments manufactured. This data is not guaranteed, does not include measurement uncertainty, and is valid only at room temperature (standardized to 25 °C).

Nominal values (nom)

Characterizes product performance by means of average performance of a representative value for the given parameter (e.g. nominal impedance). This data is not guaranteed and is valid only at room temperature (standardized to 25 °C).

Measured values (meas) Characterizes expected product performance by means of measurement results gained from individual samples.

Specifications are subject to change without notice. For the most recent product specifications, visit www.signalcore.com.

Spectral Specifications

RF input range (1)	1 MHz to 3.9 GHz
IF output center frequency	70 MHz
IF output polarity (2)	Non inverted/Inverted
IF bandwidth (3 dB)	
Final IF filter bypassed	> 20 MHz
Final IF filter enabled (default)	> 18 MHz

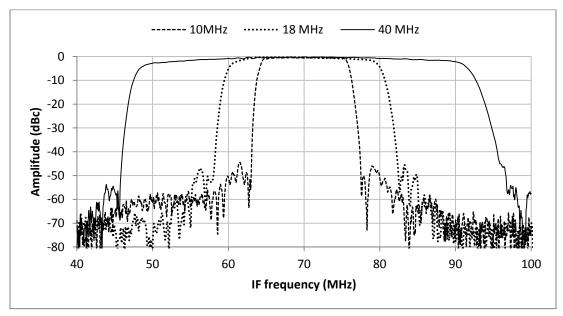


Figure 1. Typical output IF response of filter options; two IF filter options are available. Standard IF option has a 20 MHz bandwidth path and a bypass (no filter) path.

- (1) RF input below 1 MHz suffers from amplitude roll-off and calibration is not valid below this lower-end frequency. In the frequency range below the specified IF bandwidth (< ~15 MHz) the first LO leakage appears inside the IF band. This LO leakage will appear as ~DC when the RF is converted to baseband in the final analysis. Furthermore, because the LO appears inside the IF band it will inter-modulate with the input RF signal to produce higher order in-band spurious signals that may degrade signal integrity. It is recommended to attenuate the RF signal before the first mixer by applying RF attenuation or attenuate after the first mixer by applying IF1 attenuation. Suppressing the RF amplitude in front of the downconversion path will reduce the spurious signal levels.
- (2) The IF output polarity refers to the conversion polarity of the downconverter. When the polarity is inverted, the spectral content of the output is inverted with respect to the input; this process is commonly known as "spectral inversion" or "spectral flipping". The choice depends on the application. For digitizers that are sampling the IF in the even order Nyquist zones that naturally inverting spectra, having the IF polarity inverted will produce non-inverted baseband, and vice-versa. However this is only a convenience in this application case because inverted spectrum, once digitized, can easily be re-inverted mathematically.

RF tuning

Frequency step resolution (3)	1	Hz
Lock and settling times (4)	1	ms

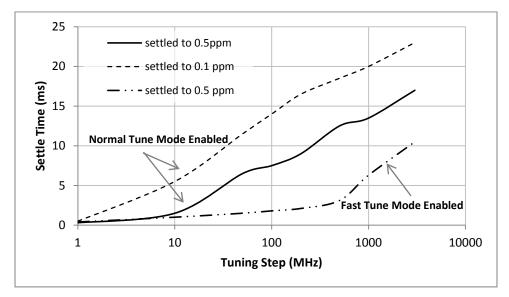


Figure 2. Typical frequency settling time versus tuning step with a 3600 MHz final frequency.

- (3) To give users flexibility, the device has three resolution modes; two coarse modes and one fine modes. The coarse modes using fractional N PLL allow 1 MHz and 50 kHz steps while the fine modes using PLL and DDS provide less than 1 Hz resolution. See the appropriate sections of this manual for further information.
- (4) Locked and settled to < 1 ppm of final frequencies of > 500 MHz and step size of < 10 MHz. For final frequencies < 500 MHz the settle time applies to accuracy with 500 Hz of the final frequency for a 10 MHz step. See Figure 7 for examples of other tuning step settling times. When fast-tune mode is enabled the noise damping capacitor across the main YIG tuning coil is disengaged, resulting in an increase of the rate of current flow through the coil and settling to a steady state quicker. Lock time begins when the full tuning word command is received by the device.

Frequency reference (5)

Technology	Temperature compensated crystal oscillator
Accuracy ± [(aging x last adjustr	ment time lapse) + temp stability + cal accuracy]
Initial calibration accuracy	±0.05 ppm
Temperature stability (6)	
20 °C to 30 °C	±0.25 ppm
0 °C to 55 °C	±1.0 ppm
Aging	± 1 ppm for first year @ 25 $^{ m oC}$

- (5) The frequency reference refers to the device's internal 10 MHz TCXO time-base. Accuracy is in parts-per-million or ppm $(1x10^{-6})$.
- (6) Users must apply sufficient cooling to the device to keep the unit temperature as read from its internal temperature sensor within the range of 40 $^{\circ}$ C to 45 $^{\circ}$ C at an ambient temperature of 25 $^{\circ}$ C.
- (7) Accuracy of the device for any given input RF signal.

Sideband phase noise (dBc/Hz) (8)(9)(11)

RF Frequency

Offset	100 MHz	1000 MHz	2000 MHz	3500 MHz
100 Hz	-88	-87	-85	-83
1 kHz	-100	-99	-98	-97
10 kHz	-108	-107	-106	-105
100 kHz	-119	-118	-117	-115
1 MHz	-143	-142	-142	-141
10 MHz	-152	-152	-150	-149

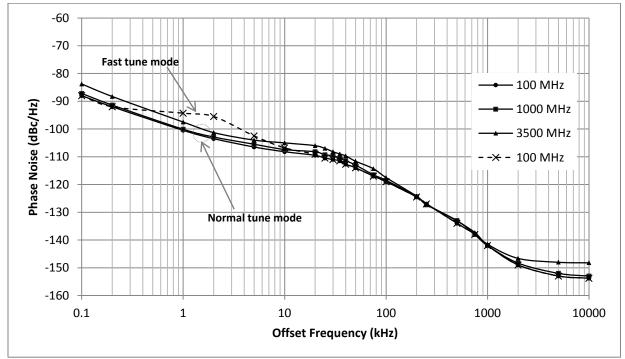


Figure 3. Typical measured sideband noise. (9)

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- (8) Sideband phase noise as specified is based on measured sideband noise which includes both phase noise and amplitude noise contributions. Sideband noise is specified for the downconverter when tune mode is set to NORMAL. See the appropriate sections in the user manual for further information on how to set the device to NORMAL or FAST-TUNE modes.
- (9) These results are obtained with input signal levels of 0 dBm at the mixer (no RF attenuation) and the output IF level set to 3 dBm. The source is an ultra-low noise 100 MHz OCXO with noise floor of -176 dBc/Hz. The 1000 MHz and 3500 MHz signals were multiplied up from the same OCXO. The floor of the multiplied up 3500 MHz signal was about -143 dBc/Hz so a phase-locked YIG oscillator was used to complete the measurement for offset frequencies greater than 500 kHz. The YIG oscillator noise floor was better than -160 dBc/Hz. In FAST-TUNE mode the noise damping capacitor across the YIG tuning coil is disengaged, and as a result the close-in phase noise degrades.

LO related sideband spurious signals (10)(11)

< 200 kHz	-75	dBc
> 200 kHz	-80	dBc

- (10) Sideband spurious signals are usually the result of the local oscillators in the device. Sources of sideband spurious signals in the synthesized local oscillators are primarily due to fractional-N spurious products in the PLL's, DDS noise sources, and intermodulation between oscillators within the multiple-loop PLL synthesizers. Fractional-N and DDS spurious products affect spectral region below 200 kHz and intermodulation products affect spectral regions out to a couple of MHz. SignalCore uses mathematical algorithms to properly select the synthesizer parameters used in the multiple-loop fractional-N PLL to ensure that typical sideband spurious products are better than the specifications.
- (11) Specifications are valid for all modes of frequency tuning, whether is it PLL only mode or DDS driven mode. As the YIG oscillator is sensitive to magnetic fields, magnetic noise due to electrical fans, supply transformers, and other magnetic field-producing devices may induce sideband noise on the signals when they are placed in close proximity. It is recommended that users exercise good technical judgment when such accessories are needed (e.g., mounting a cooling fan directly onto the RF enclosure of the device.

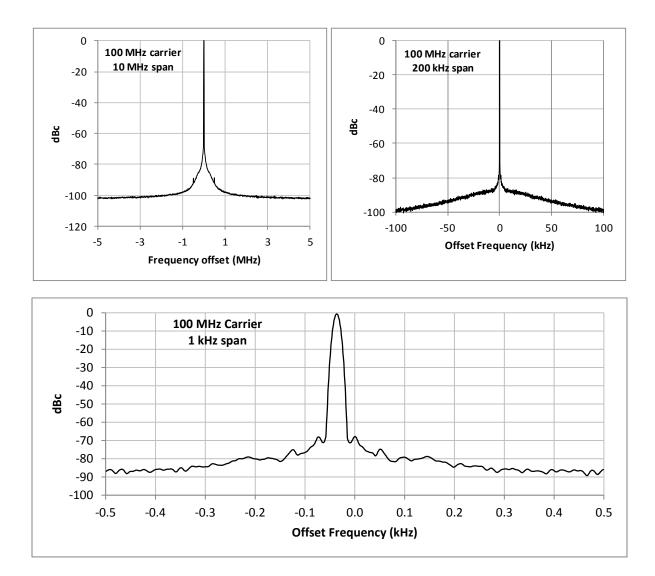


Figure 4. Plots show the raw spectral purity for a 100 MHz input RF signal (LO = 4.775 GHz). Note that the power supply noise of 60 Hz and its harmonics are in the noise. The measurement instrument is not phase-locked to the unit under test.

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Amplitude Specifications

Input range	
AC (preamplifier disabled)	
AC (preamplifier enabled)	
DC (12)	0 V
(12) Large and fast DC transients could damage the input solito 10 V is sustainable.	id state devices. Slow ramp up of DC
Attenuation range	
RF	0 to 30 in 1 dB steps
IF (13)	0 to 90 in 1 dB steps
Input voltage standing wave ratio (VSWR)	
Preamp off, 0 dB input attenuation	
10 MHz to 2.4 GHz	< 1.5
2.5 GHz to 3.6 GHz	< 1.8
Preamp on, 0 dB input attenuation	
10 MHz to 2.4 GHz	< 1.5
2.5 GHz to 3.6 GHz	< 2.0
Gain range (@ 1GHz) (14)	
Minimum (15)	
Maximum (preamplifier disabled) (16)	30 dB typical
Maximum (preamplifier enabled) (16)	50 dB typical
Preamplifier gain	20 dB typical
RF amplitude accuracy (15 °C to 35 °C ambient)	
RF gain response flatness (uncorrected)	8 dB typical
RF gain flatness response (corrected) (17)	±0.75 dB
Absolute gain accuracy (corrected) (17)	±0.9 dB (±0.5 dB typical)
IF Amplitude accuracy (15 °C to 35 °C ambient)	
IF in-band response flatness (uncorrected)	3 dB typical

IF in-band response (corrected) (16) ±0.5 dB

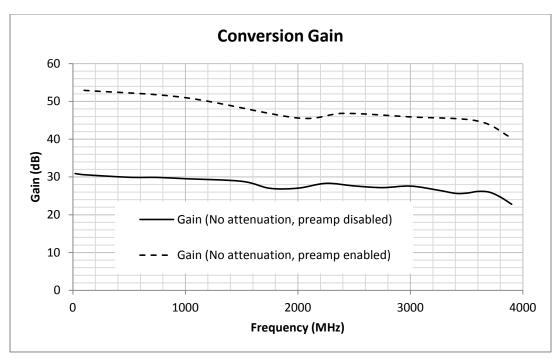


Figure 5. Typical RF conversion gain response @ 25 °C.

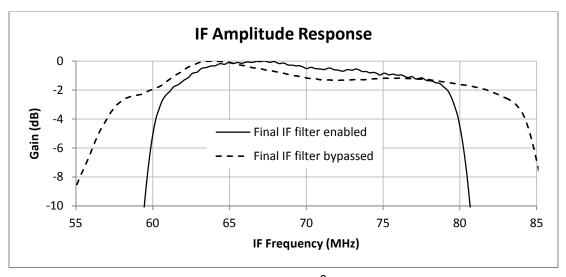


Figure 6. Typical IF amplitude response @ 25 °C for standard 20 MHz IF filter.

RF to IF group delay (80% of IF bandwidth)

Final IF filter enabled	1 μs typica
Final IF filter bypassed	100 ns typica
IF phase linearity (80% of IF bandwidth) (18)	
Final IF filter enabled	+/- 8 degree
Final IE filter hypassed	+/- 8 degree

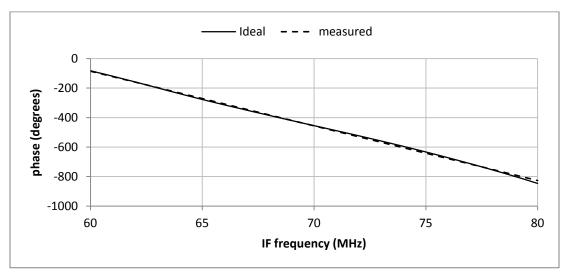


Figure 7. Phase deviation over 20 MHz.

- (13) There are three IF attenuators in total, each having 30 dB of attenuation. There are two attenuators in the final stage, and one attenuator in the first IF stage after the first mixer. How to effectively use them to optimize for performance is outlined in the appropriate sections of this manual.
- (14) These are typical gain specifications. The gain of the device is calibrated and stored in the device calibration EEPROM.
- (15) Minimal gain is specified when all attenuators are set to their maximum values and the RF pre-amplifier is disabled.
- (16) Maximum conversion gain is specified when all the attenuators are set to 0 dB attenuation.
- (17) Correction stored in the calibration EEPROM must be applied properly. Users are not obligated to use the calibration provided, may devise their own method of calibration and correction should they choose to. User methods of calibration and application may improve on the accuracies specified.
- (18) For broadband signal operation it is recommended that users apply in situ amplitude and phase equalization to the received signal to minimize amplitude and phase errors caused by the device. Phase deviation at offset frequencies from the center frequency of 70 MHz is stored in the calibration EEPROM. The calibration may be applied as a first order correction.

Dynamic Range Specifications

Spurious response (19)

Residual spurious signals (20)	< -100 dBm
LO related spurious signals (21)	< -80 dBc
Image rejection (22)	<-100 dBc
IF rejection (23)	<-115 dBc

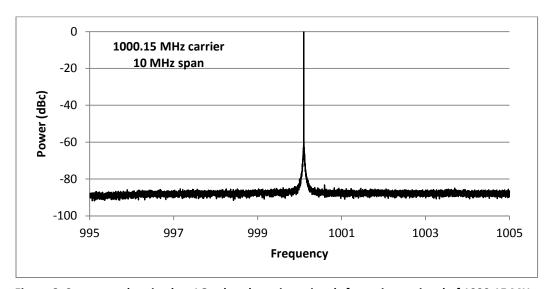


Figure 8. Spectrum showing low LO related spurious signals for an input signal of 1000.15 MHz.

- (19) Spurious responses are unwanted signals appearing at the IF output. All spurious products are referenced to the RF input, meaning that they are treated as if they originate at the input port of the device.
- (20) Residual spurious signals are observed and referenced to the RF input of the device when the RF input is terminated with a matched load. The RF and first IF (IF1) attenuators are set to 0 dB attenuation and the final IF attenuators were adjusted to obtain an overall device gain of 20 dB. The preamplifier is disabled.
- (21) LO related spurious signals are unwanted signals produced at the IF output due to intermodulation of the local oscillators. These spurious signals are measured relative to an RF signal present at the input. The specification referenced here is for a device configuration of -20 dBm at the mixer, 0 dBm at the IF output, and a total gain of 20 dB.
- (22) Image rejection is the ability of the device to reject an image signal of the RF frequency that would otherwise produce the same result as the desired RF signal. The image of the desired RF signal is calculated as $RF_{image} = RF + 2IF_1$, where $IF_1 = 4.675 \ GHz$.
- (23) IF rejection is the ability of the device to reject RF signals at any of the IF frequencies while the device is tuned elsewhere. Signal level at the mixer is -20 dBm and total gain is 20 dB.

Input noise (15 °C to 35 °C ambient) (24)

Preamplifier disabled (25)

	100 MHz	1000 MHz	3600 MHz
Noise floor (dBm/Hz)	-153	-152	-148
Noise figure (dB)	21	22	26

Preamplifier enabled (25)

	100 MHz	1000 MHz	3600 MHz
Noise floor (dBm/Hz)	-167	-166	-164
Noise figure (dB)	7	8	10

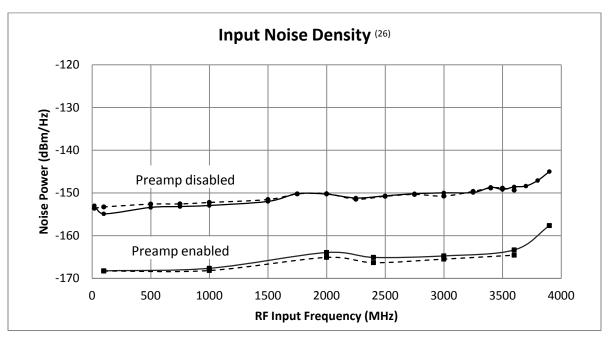
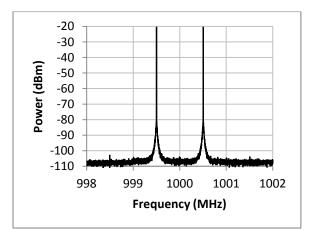


Figure 9. Measured noise density of the average of two lots.

- (24) Noise (thermal) is referred to the input of the device.
- (25) The device is configured with 0 dB RF attenuation, 0 dB IF1 attenuation, and IF attenuators adjusted to set the gain to 20 dB. This setting is made to be consistent with the configuration for other specifications such as linearity and spurious responses so that the user may obtain a clearer picture of the specified performance of the device. The RF input is terminated with a matched 50 Ω load.
- (26) In spectrum analyzer and signal analyzer applications this is also commonly referred to as the Displayed Average Noise Level (DANL). This assumes that the digitizer used does not limit the performance of the device.

Input third-order intermodulation (IIP3, dBm)

	100 MHz - 1 GHz	1 GHz - 2.5 GHz	2.5 GHz - 3.9 GHz
Preamplifier disabled (27)(29)	16 [17]	17.5 [20]	18.5 [20]
Preamplifier enabled (28)(29)	-5.0 [-2]	-3.0 [-1]	-2.0 [0]



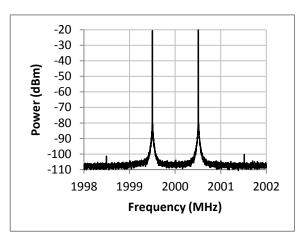


Figure 10. Plots show the typical IMD performance with two -20 dBm signals at the input, 0 dB RF attenuation, preamp disabled, and conversion gain of 20 dB.

- (3) Specifications are based on 0 dB RF attenuation, 0 dB IF1 attenuation, two -20 dBm tones with 1 MHz separation at the mixer, and final IF attenuators set to maintain 0 dBm at the IF output.
- (4) Specifications are based on 0 dB RF attenuation, 0 dB IF1 attenuation, two -30 dBm tones with 1 MHz separation at the mixer, and final IF attenuators set to maintain 0 dBm at the IF output.
- (5) These are in-band measurements and not out-of-band measurements. Out-of-band signal tones exist outside the IF filter bandwidth of the device, and thus may provide better IP3 measurements. However, using in-band signal tones provides better estimation of the device's non-linear effects on broadband signals.

Input second harmonic distortion (SHI, dBm)

Input second harmonic intercept point (dBm)	400 MHz	1000 MHz	1.8 GHz
Preamplifier disabled	62	62	58
Preamplifier enabled	32	33	30

Input compression point (dBm)

	100 MHz - 1 GHz	1 GHz - 2.5 GHz	2.5 GHz - 3.9 GHz
Preamplifier disabled	1	1.5	2
Preamplifier enabled	-23	-20	-19

Dynamic range

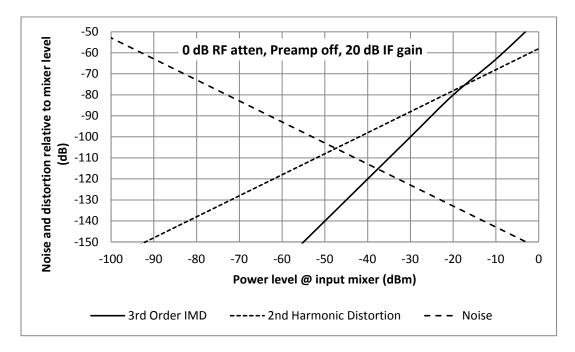


Figure 11. Instantaneous dynamic ranges plotted with preamplifier disabled for 1000 MHz measured data. Mixer level is at input level.

- (30) Measurement dynamic range refers to the device SNR measurement capability using two or more configurations settings. For example, the user could set sufficient RF attenuation to capture the high level signals and then turn on the preamplifier to measure low level noise.
- (31) Instantaneous dynamic range refers to the instantaneous device SNR measurement using a single configuration setting. For example, the user could set the downconverter to receive a 0 dBm signal at the mixer, while at the same setting be able to measure the signal noise floor to -150 dB below its peak.

Reference Input and Output Specifications

Reference output specifications

Center frequency (32)	10 MHz/100 MHz
Amplitude	6 dBm min typ
Waveform	Sine
Impedance	50Ω nominal
Coupling	AC
Connector type	SMA female
Frequency accuracy	See "Spectral Specifications" section

Reference input specifications

Center frequency	10 MHz
Amplitude	10 dBm min/ +13 dBm max
Phase-lock range	± 10 ppm (typ)
Impedance	50 Ω nominal
Coupling	AC
Connector type	SMA female

(32) The output reference frequency may be selected programmatically for 10 MHz or 100 MHz. The 100 MHz reference may be used to drive a digitizing ADC directly. Refer to the **"Frequency reference"** specifications under "Spectral Specifications" for frequency accuracy.

Port Specifications

RF input

Input impedance	50 Ω
Coupling	AC
Connector type	SMA female
LO leakage	

IF output

Output impedance	50 Ω
VSWR	1.6
Coupling	AC
Connector type	SMA female
Output amplitude	

General Specifications

Environmental

Op	perating temperature (1)	0 °C to +40 °C
Sto	orage temperature (2)	40 °C to +70 °C
Op	perating relative humidity (3)	10% to 90%, non-condensing
Sto	orage relative humidity (4)	5% to 90%, non-condensing
Op	erating shock (5)	
Sto	orage shock ⁽⁶⁾	50 g, half-sine pulse, 11 ms duration
Ор	perating vibration ⁽⁷⁾	5 Hz to 500 Hz, 0.31 g _{rms}
Sto	orage vibration ⁽⁸⁾	5 Hz to 500 Hz, 2.46 g _{rms}
Alt	titude 2,000 m maximum (mair	ntaining 25 °C maximum ambient temperature)
Physic	cal	
	Dimensions (W x H x D, max envelope)	3.7" x 1.4" x 6.1"
	Weight	2.6 lbs
	Input voltage	12 VDC
	Power consumption	34 W typical
	Communication interface	USB and RS-232/ SPI
Safety	⁷	Designed to meet the requirements of:
	IEC 6	1010-1, EN 61010-1, UL 61010-1, CSA 61010-1
Electr	omagnetic Compatibility (EMC)	Designed to meet the requirements of:
		ons; Basic immunity 1, EN 55011 (CISPR 11) 1: Group 1, Class A emissions, FCC 47 CFR Part issions
CE		
	2006/95/EC; Low-Voltage Directive (safety), Directive (EMC Directive)	, 2004/108/EC; Electromagnetic Compatibility
Warra	anty 1 year parts and	d labor on defects in materials or workmanship

- (33) Meets requirements of IEC-60068-2-1 and IEC-60068-2-2. Operating temperature may be extended to +55 °C with appropriate user-provided cooling solution. Contact SignalCore for recommended minimum airflow rates.
- (34) Meets requirements of IEC-60068-2-1 and IEC-60068-2-2.
- (35) Meets requirements of IEC-60068-2-56 and MIL-PRF-28800F, Class 3.
- (36) Meets requirements of IEC-60068-2-27 and MIL-PRF-28800F, Class 3.
- (37) Meets requirements of IEC-60068-2-64 and MIL-PRF-28800F, Class 3.



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